

Robust Reception for a Satellite-based Global Rescue System

European GNU Radio Days 2023

Felix Artmann, Marcus Müller



- Introduction
 - COSPAS-SARSAT
- Techniques and Algorithms
 - Signal Detection: Time and Phase
 - Signal Detection: Frequency Offset
 - Despreading
- Evaluation Over-the-air
- Conclusion & Outlook

COSPAS-SARSAT

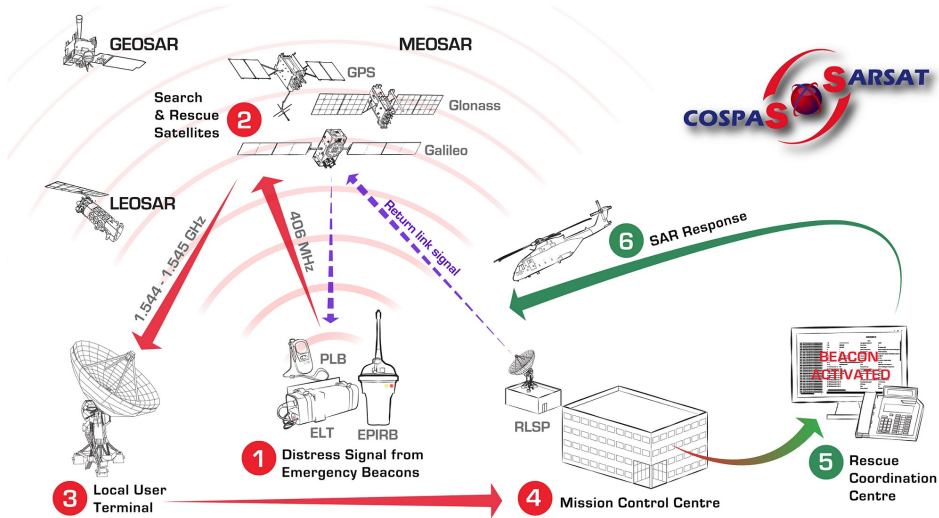


Image source: INTRODUCTION TO THE COSPAS-SARSAT SYSTEM, C/S G.003, Issue 7, November 2019



- Joint development effort with Becker AG
 - Manufacturer of aeronautical equipment
 - Had gen. 1 COSPAS-SARSAT devices, but no way to support gen. 2
- Project coordination: German Aerospace Center (DLR), financial support bei German federal Ministry of Economy Affairs and Climate Action (BMWK)
- CEL Goals:
 - Assist Becker in transition to SDR
 - Develop reference implementation
 - Value-add functionality through overlay system and awareness
- CEL Theses:
 - **Reference transceiver** (Elena Schütz, Felix Artmann)
 - Overlay in satellite uplink (Messmer)
 - Overlay in auxiliary chirp homing signal (Gölz, Manz, Messmer, . . .)

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- Information bit stream: 300 bit s^{-1}
 - Message length = 300 bit
 - Preamble length = 50 bit
- **Offset-QPSK**
- **Direct Sequence Spread Spectrum**
 - Spread using Pseudo Random Noise (PRN)
 - Two different sections of an **m-Sequence** for in-phase and quadrature components
 - **I** : PRN₁
 - **Q**: PRN₂
- **Properties**
 - Spread factor $L = 256$
 - Output stream: 38400 chips/second
 - Signal hidden from others

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■ Evaluation Over-the-air

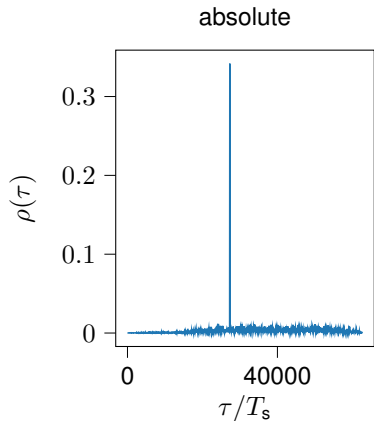
■ Conclusion & Outlook

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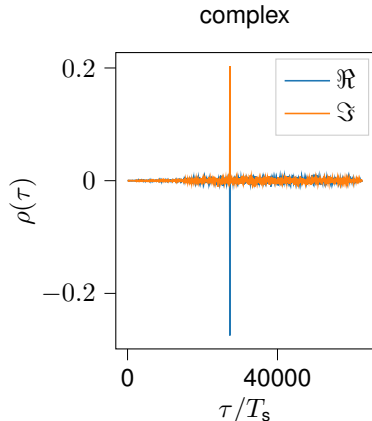
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(a) Time



(b) Phase

- Correlation with preamble (6400 chips)
- Sent signal

$$\Re\{A_{Tx}\} = \text{PRN}_1$$

$$\Im\{A_{Tx}\} = \text{PRN}_2$$

- Received signal

$$\Re\{A_{Rx}\} = \Theta_1 \cdot \text{PRN}_1 + \Theta_2 \cdot \text{PRN}_2$$

$$\Im\{A_{Rx}\} = \Theta_2 \cdot \text{PRN}_1 - \Theta_1 \cdot \text{PRN}_2$$

$$A_{\text{corr}} = A_{Rx} \cdot e^{-j \cdot \varphi_{\text{peak}}}$$

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■ Techniques and Algorithms

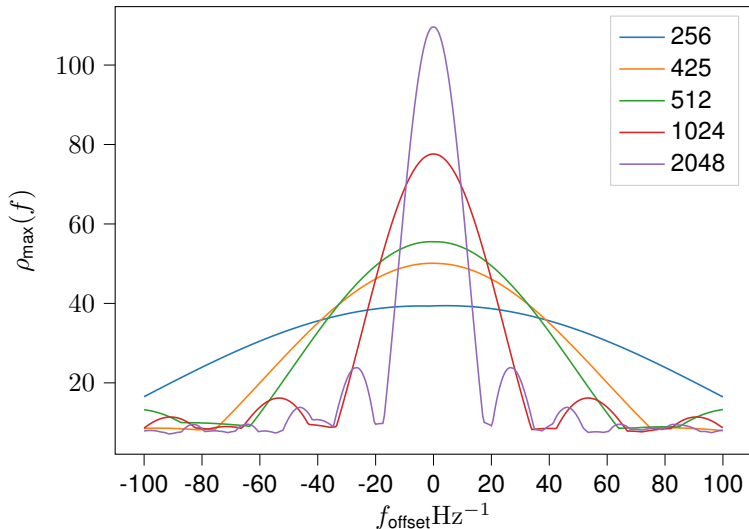
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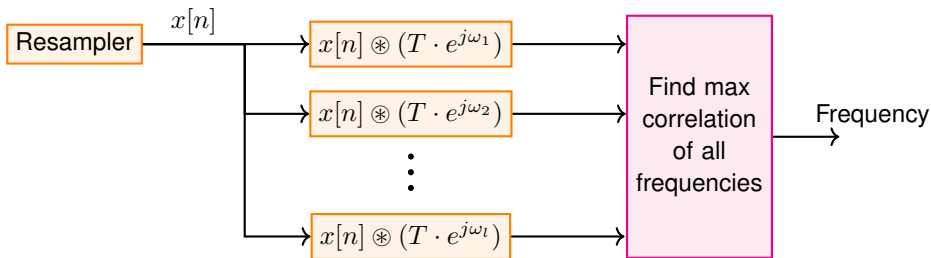
Frequency Offset

- Same procedure as on slide before but **truncated** preamble
- Deviation expected in the range of at least -400 Hz to 400 Hz
- Peak gets barely visible
- Need for **Frequency Detection**



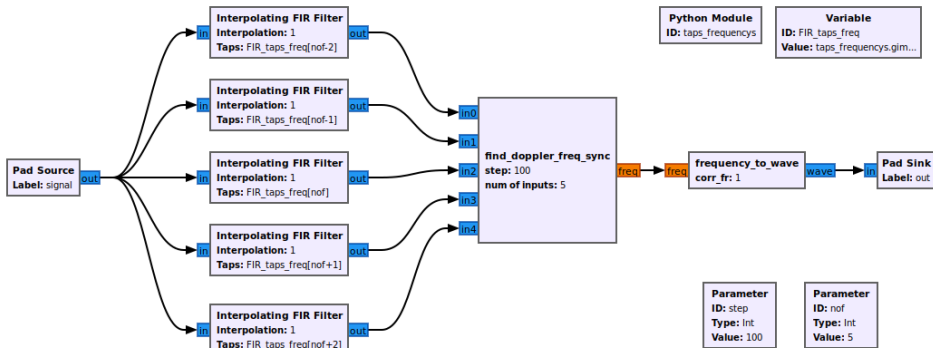
Signal Detection: Correlator Bank

- Correlation with several frequency shifted versions of the preamble (T)
- Eventually reducing the susceptibility for frequency variations by only using fraction of the preamble
 - By that: possible to increase $\Delta\omega$



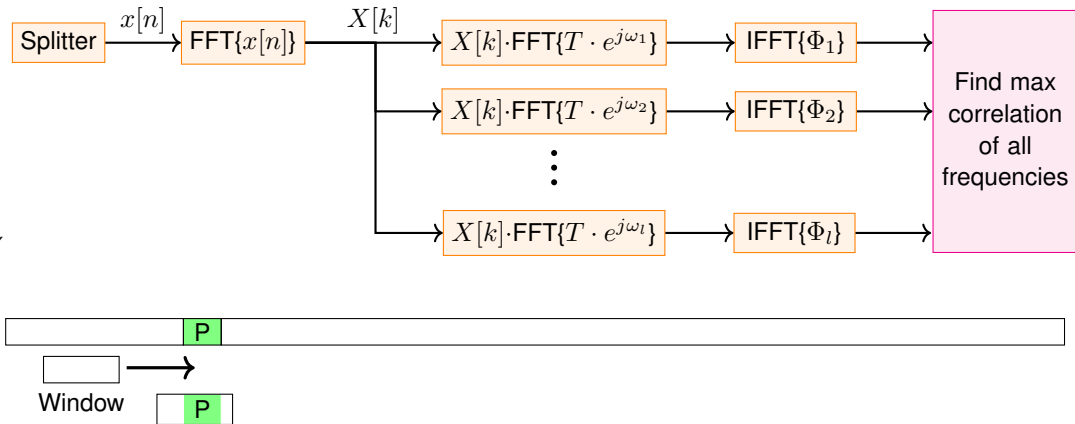
Signal Detection: Correlator Bank – GRC

- Using FIR filters
- First stage; $f_{\text{grid}} = 100 \text{ Hz}$; covered spectrum: $-250 \text{ Hz} < f < 250 \text{ Hz}$

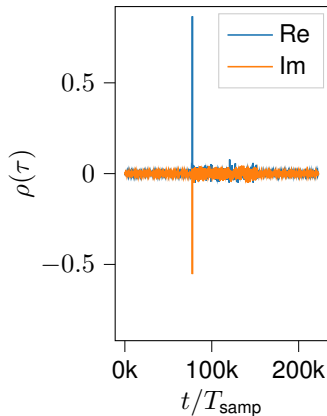


Signal Detection: Filters in DFT Domain

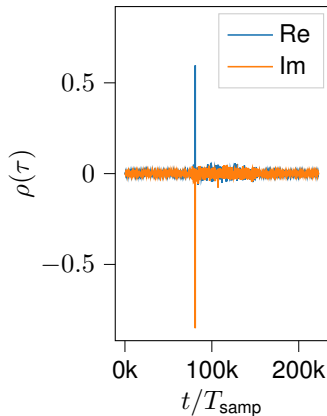
- Use Fast Fourier Transform



Fine Frequency Sync: Split Preamble Method



(a) first half of preamble



(b) second half of preamble

■ **So far:** Coarse estimation of frequency

■ **Now:** Exact determination by evaluating the phase shift

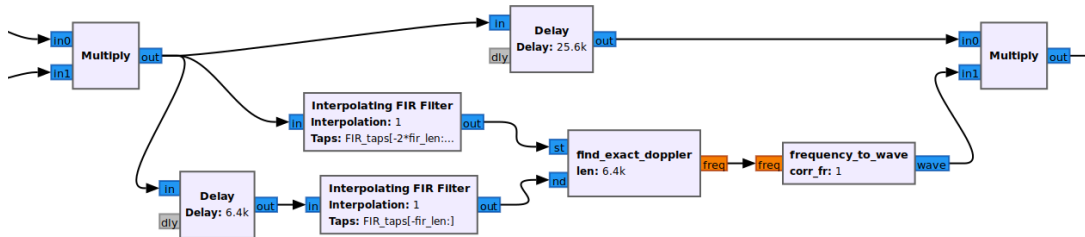
- Split preamble
- Difference of phase angles of correlation peaks

$$\varphi_{\text{diff}} = \varphi_B - \varphi_A \quad (1)$$

$$t_{\text{delay}} = \frac{3200}{38400 \frac{1}{s}} = \frac{1}{12} s = 0.08333s \quad (2)$$

$$f_{\text{diff}} = \frac{\varphi_{\text{diff}}}{t_{\text{delay}} \cdot 2 \cdot \pi} \quad (3)$$

Fine Frequency Sync: Split Preamble Method (GRC)



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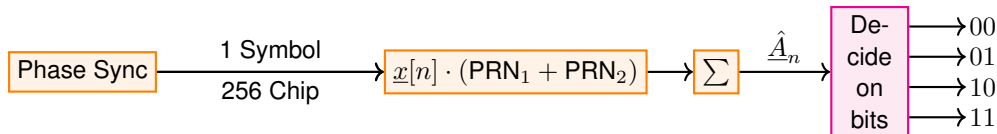
■ Techniques and Algorithms

- Signal Detection: Time and Phase
- Signal Detection: Frequency Offset
- **Despreading**

■ Evaluation Over-the-air

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Despreading: Σ -PRN



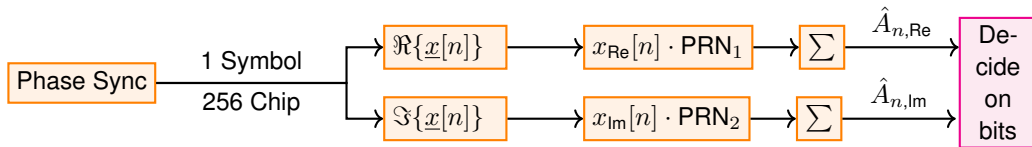
- Despreading by multiplication of BPSK-symbols with BPSK-preamble
- Computationally advantageous
- Quadrature cross-talk if phase not perfect

$$\underline{x}[n] = a[n] + jb[n]$$

$$P[n] = a[n] \cdot (\text{PRN}_1 + \text{PRN}_2) + jb[n] \cdot (\text{PRN}_2 + \text{PRN}_1)$$

$$\lim_{\text{SNR} \rightarrow \infty} P \approx \sum_{n=1}^{256} a[n] \cdot \text{PRN}_1 + jb[n] \cdot \text{PRN}_2$$

Despreading: Split-IQ PRNs

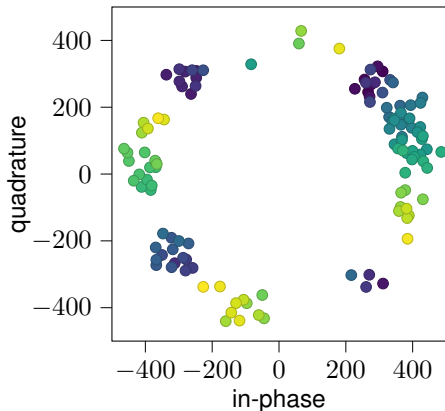


- Despreading separately for PRN₁ and PRN₂
- Two multiplications in real domain
- Eliminates quadrature cross-talk, halves noise power

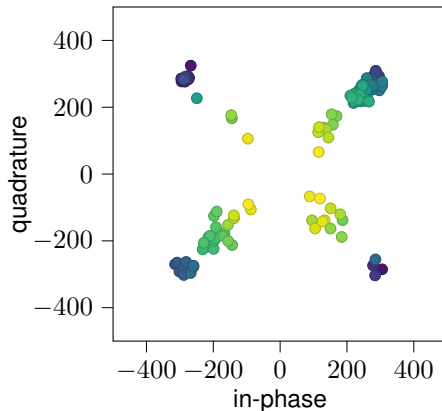
Despreading: Comparison

- Frequency error can change after preamble
- Compare constellation diagrams with error of 0.3 Hz

blue $\xrightarrow{\text{time}}$ yellow



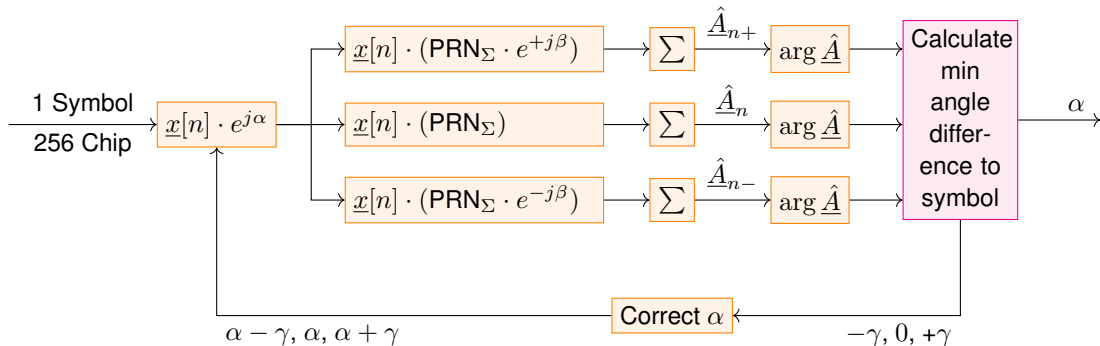
(a) Σ -PRN despreading



(b) Split-IQ despreading

Adaptive Despreading: Phase Error Feedback

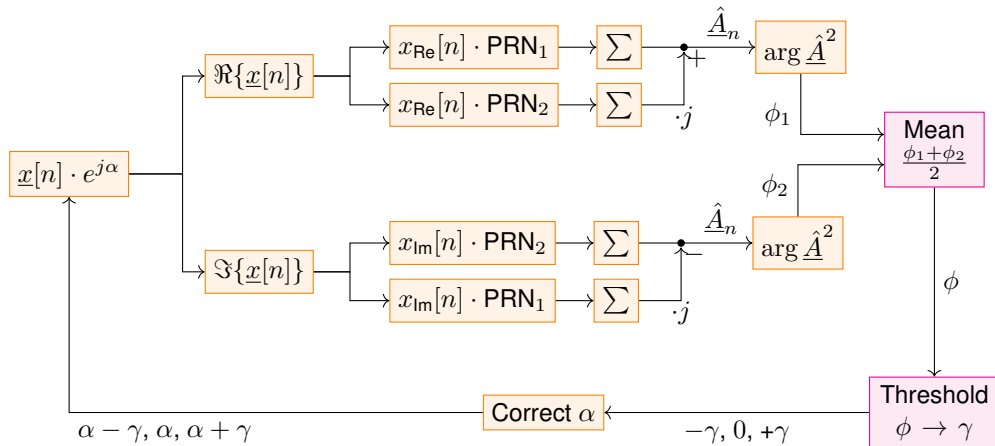
- Despreading with feedback of phase error based on Σ -PRN method
- $\text{PRN}_{\Sigma} = \text{PRN}_1 + \text{PRN}_2$



Problem: Still using the worse sum-PRN for despreading.

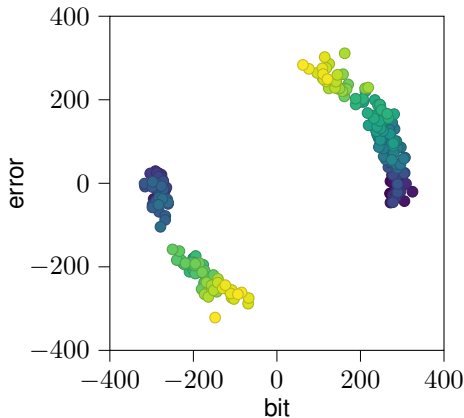
Adaptive Despreading: Split-IQ with Phase

- Method to use Split-IQ and still correct evolving phase error?

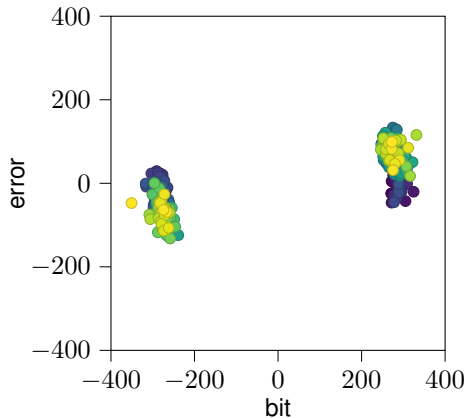


Adaptive Despreading: Results

Simulated frequency error of $f_{\text{off}} = -0.3 \text{ Hz}$

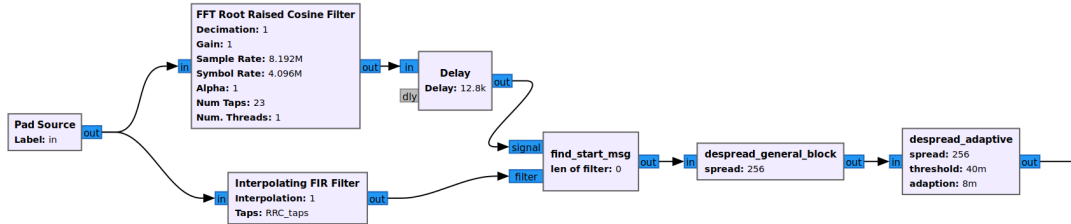


(a) without correction



(b) correction w. $\phi_{\text{thres}} = 2\pi \cdot 0.05$ and $\gamma = 2\pi \cdot 0.01$

- Adaptive Despreading also implemented



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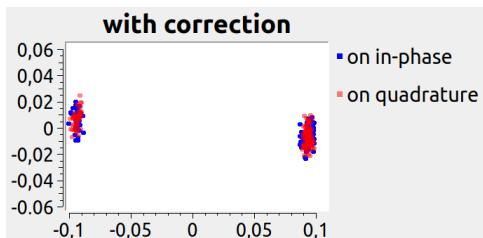
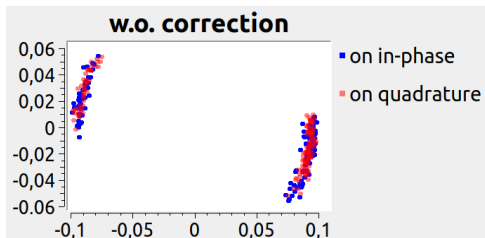
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Over-the-air results

- Tests using two different USRPs over the air
- Detected frequency offset around 100 Hz
- Detection and synchronization **successful**

- Adaptive despread
- Constellation diagrams from GNU Radio GUI with real-world data and identical stream



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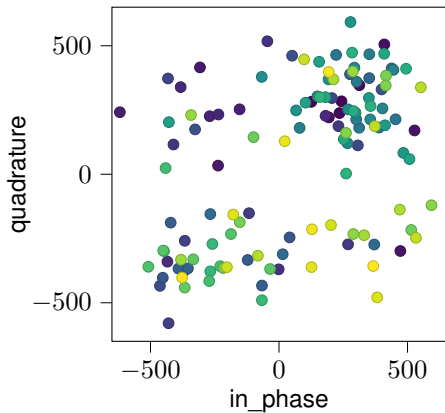
Take-Away

- Creation of (as far as we know) first functional transceiver for 2nd gen COSPAS-SARSAT
- Usage of improved frequency correction methods pushes beyond standard
- Over-the-air proven, suitable for deeply embedded implementation

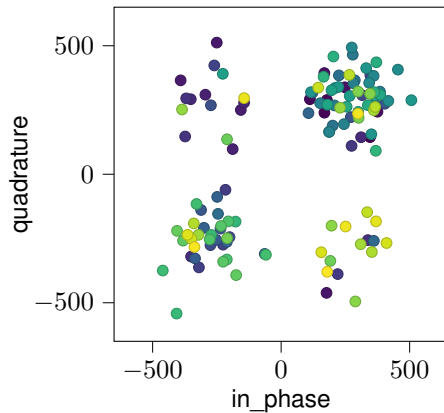
Next up

- Integration with overlay systems
- GNU Radio through MEO
- Shrink-Down to microcontroller (Becker)

Questions?



(a) Σ -PRN method



(b) Split-IQ method

- Using **GNU RADIO**
- Communication via USRPs



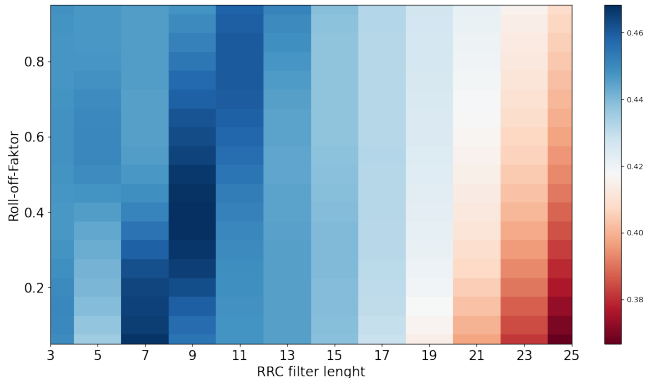
- **GNU RADIO** is a Toolkit

Image source: <https://www.gnuradio.org/>

Optimization of RRC Filter Length and Roll-off-Factor

- Simulation of different output and mf filters
- Normalization of power
- Adding AWGN
- Compare correlation¹

- Optimum at around $\beta = 0.35$ and $N = 9$ Taps²

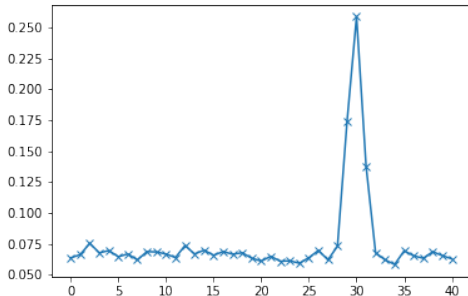


¹The result is not yet deterministic as it includes randomness

²Simulation with SNR=-10dB

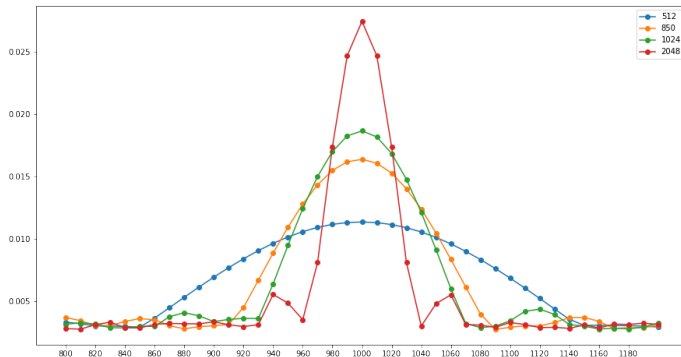
Doppler Shift Correction

- What can be done?
 - Correlation with several frequency shifted versions of the preamble
 - Reducing the susceptibility for frequency variations by only using fraction of the preamble (next slide)
- Using only 512 chips of preamble
- Vary over a range of -2000 Hz to 2000 Hz
- Raster of 100 Hz



Optimization of Doppler Shift Correction

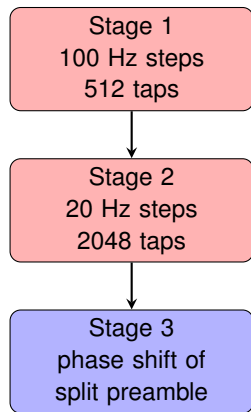
- By shortening the length of the correlated preamble frequency sensitivity gets smaller but peak also gets lower
 - Frequency raster can be higher
 - **But:** Risk of not finding the peak
- Simulation of different lengths



Implementation of Receiver (Previous Version)

- Which method will be used?
- Consideration of least possible resource usage but also reliable detection

- Use three-stage doppler correction
 - Frequency-shifted preamble taps over whole spectrum³: grid of 100 Hz; 512 taps
 - Frequency-shifted preamble taps over narrow range: grid of 20 Hz; 2048 taps
 - Time shifted preamble taps with accurate detection detection by phase shift calculation



³expected doppler range from -400Hz to 400Hz